

EE3700 : Communications I

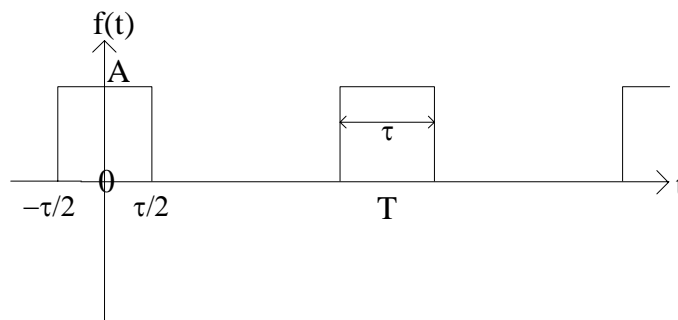
Tutorial 1- Solution

1. A certain periodic band-limited signal consists of frequencies at dc, 500Hz and at 1kHz. The signal can be written as

$$s(t) = 12 + 20 \cos 1000\pi t - 20 \sin 1000\pi t + 12 \cos 2000\pi t - 6 \sin 2000\pi t$$

Express this signal as a Fourier series in complex exponential form.

2. Determine the Fourier coefficients C_n for the following periodic signal.



3. Considering the waveform in question 2, estimate the amount of the original signals power which would pass through a filter that rejects harmonics above 2nd order (ie greater than $2 \times f_0 = 2/T$). Assume $T=1\text{mS}$, $A = 10$ volts and $\tau = 0.25\text{mS}$.
4. A radar system transmits pulses of 200nS duration. Estimate the bandwidth of the transmitter circuitry necessary to ensure that at least 90% of the energy of the pulse is transmitted.

Q1. The signal consists of cos and sin terms. Let's examine the complex Fourier series representation of these signals then.

COS TERMS. $x(t) = A \cos \omega_0 t$.

$$\begin{aligned}
 C_n &= \frac{A}{T} \int_0^T \cos \omega_0 t e^{-jn\omega_0 t} dt \quad ; n=1,2,3,\dots \\
 &= \frac{A}{T} \int_0^T \cos \omega_0 t [\cos n\omega_0 t - j \sin n\omega_0 t] dt. \\
 &= \frac{A}{2T} \int_0^T \cos [\omega_0 t - n\omega_0 t] + \cos [\omega_0 t + n\omega_0 t] \\
 &\quad - j \{ \sin(\omega_0 t + n\omega_0 t) - \sin(\omega_0 t - n\omega_0 t) \} dt.
 \end{aligned}$$

Note as all terms are sinusoidal with a frequency which is a multiple of ω_0 , the integrations will be zero, except when $n = \pm 1$.

$$\therefore C_1 = \frac{A}{2} \quad \& \quad C_{-1} = \frac{A}{2}.$$

$$\begin{aligned}
 \text{AND } x(t) &= \sum_{n=-\infty}^{\infty} C_n e^{jn\omega_0 t} \\
 &= \frac{A}{2} e^{j\omega_0 t} + \frac{A}{2} e^{-j\omega_0 t}
 \end{aligned}$$

SIN TERMS $x(t) = A \sin \omega_0 t$.

$$\begin{aligned}
 C_n &= \frac{A}{T} \int_0^T \sin \omega_0 t [\cos n\omega_0 t - j \sin n\omega_0 t] dt. \\
 &= \frac{A}{2T} \int_0^T \sin \omega_0 t (1+n) + \sin \omega_0 t (1-n) \\
 &\quad - j \{ \cos \omega_0 t (1-n) - \cos \omega_0 t (1+n) \} dt.
 \end{aligned}$$

Again all terms, except where $n = \pm 1$, will be zero.

$$C_1 = -j \frac{A}{2} \quad C_{-1} = +j \frac{A}{2}$$

$$s(t) = -j \frac{A}{2} e^{j\omega_0 t} + j \frac{A}{2} e^{-j\omega_0 t}$$

In question 1). $s(t) = 12 + 20 \cos 1000\pi t - 20 \sin 1000\pi t + 12 \cos 2000\pi t - 6 \sin 2000\pi t$

Set $\omega_0 = 1000\pi$ rad/s.

$$s(t) = 12 + 20 \cos \omega_0 t - 20 \sin \omega_0 t + 12 \cos 2\omega_0 t - 6 \sin 2\omega_0 t$$

$$C_0 = \frac{1}{T} \int_0^T s(t) dt \text{ is the AVERAGE of } s(t)$$

$$= 12.$$

$$\begin{aligned} s(t) = 12 + \frac{20}{2} e^{j\omega_0 t} + \frac{20}{2} e^{-j\omega_0 t} + \frac{20j}{2} e^{j\omega_0 t} - \frac{20j}{2} e^{-j\omega_0 t} \\ + \frac{12}{2} e^{j2\omega_0 t} + \frac{12}{2} e^{-j2\omega_0 t} + \frac{6j}{2} e^{j2\omega_0 t} - \frac{6j}{2} e^{-j2\omega_0 t} \end{aligned}$$

$$\begin{aligned} s(t) = 12 + 10 e^{j\omega_0 t} + 10 e^{-j\omega_0 t} + j10 e^{j\omega_0 t} - j10 e^{-j\omega_0 t} \\ + 6 e^{j2\omega_0 t} + 6 e^{-j2\omega_0 t} + j3 e^{j2\omega_0 t} - j3 e^{-j2\omega_0 t} \end{aligned}$$

OR

$$s(t) = 12 + (10 + j10) e^{j\omega_0 t} + (10 - j10) e^{-j\omega_0 t} + (6 + j3) e^{j2\omega_0 t} + (6 - j3) e^{-j2\omega_0 t}$$

As a line spectra the signal would have spikes at dc, ω_0 and $2\omega_0$. The magnitudes are given from the expression above. eg. At $\omega_0 |C_1| = \sqrt{200}$

No 2. These can be a bit tricky! The integration can be difficult.

We can integrate our function over any period, T . In this case lets integrate from $-T/2$ to $T/2$.

$$\begin{aligned} C_n &= \frac{1}{T} \int_{-T/2}^{T/2} f(t) e^{-jn\omega_0 t} dt. \\ &= \frac{1}{T} \int_{-T/2}^{T/2} A e^{-jn\omega_0 t} dt. \\ &= \frac{A}{T} \left[\frac{e^{-jn\omega_0 t}}{-jn\omega_0} \right]_{-T/2}^{T/2} = \frac{A}{T} \left[\frac{e^{-jn\omega_0 T/2}}{-jn\omega_0} - \frac{e^{jn\omega_0 T/2}}{-jn\omega_0} \right] \\ &= \frac{2A}{-n\omega_0 T} \left[\frac{e^{j(-n\omega_0 T/2)} - e^{-j(-n\omega_0 T/2)}}{2j} \right] \end{aligned}$$

$$C_n = \frac{2A}{jn\omega_0 T} \sin(+n\omega_0 T/2)$$

$$C_0 = \text{Average Voltage} = A \cdot \frac{T}{T}$$

No3 Firstly, calculate the power of the signals from their time response. When talking about power we need to use RMS. values of voltage though!

That is;

$$\text{POWER} = [V_{\text{RMS}}]^2.$$

$$\text{where } V_{\text{RMS}} = \sqrt{\frac{1}{T} \int_{-T/2}^{T/2} f(t)^2 dt} \quad \left\{ \begin{array}{l} \text{is the square root of the} \\ \text{mean of } f(t) \text{ squared.} \end{array} \right.$$

$$b) \quad V_{\text{RMS}} = \sqrt{\frac{1}{T} \int_{-T/2}^{T/2} A^2 dt} = A \sqrt{\frac{T}{T}}$$

$$\text{If } T = 1\text{ms}, A = 10\text{V} \text{ \& } \tau = 0.25\text{ms}.$$

$$V_{\text{RMS}} = \frac{10 \cdot 1}{2} = 5\text{V}$$

$$\text{Power} = \underline{\underline{25 \text{ watts.}}}$$

If we filter the signals so that only the first 2 harmonics of the original signal are allowed to pass.

$$\text{Filtered power} = \sum_{n=-2}^{n=2} |C_n|^2$$

If all the power were to pass, the sum would be from $n = -\infty$ to $+\infty$.

$$\text{Here } C_0 = A \frac{\tau}{T} \quad \& \quad C_n = \frac{2A}{n\omega\tau} \sin(n\omega\tau/2)$$

g.f. $A=10$, $\tau=0.25\text{ms}$, $T=1\text{ms}$, $\omega_0=6.283 \times 10^3 \text{ rad/sec}$

$n=0$ $C_0 = 2.5$

$$\begin{aligned} n=1 \quad C_1 &= \frac{2 \times 10}{6.283 \times 10^3 \times 1 \times 10^{-3}} \sin \left[\frac{6.283 \times 10^3 \times 0.25 \times 10^{-3}}{2} \right] \\ &= 3.183 \times \sin(0.785) = 3.183 \times 0.7068 \\ &= 2.25 \end{aligned}$$

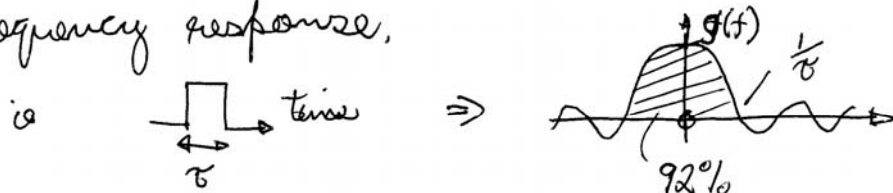
$$\begin{aligned} n=2 \quad C_2 &= \frac{2A}{2\omega_0 T} \cdot \sin \left(2\omega_0 T/2 \right) \\ &= \frac{3.183}{2} \cdot \sin(2 \times 0.785) = 1.591 \times 1.0. \\ &= 1.591 \end{aligned}$$

o.o FILTERED POWER.

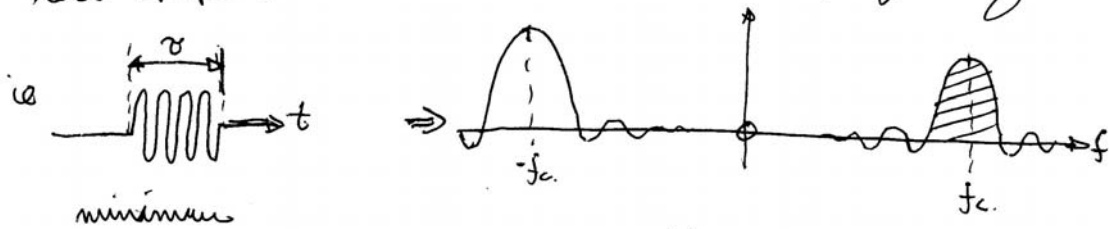
$$\begin{aligned} &= |C_{-2}|^2 + |C_1|^2 + |C_0|^2 + |C_1|^2 + |C_2|^2 \\ &= 2.53 + 5.0625 + 6.25 + 5.0625 + 2.53 \\ &= \underline{\underline{21.43 \text{ Watts}}} \end{aligned}$$

Note in this case most of the original signal power (25W) will pass through the filter. This filter is therefore reasonably efficient in this situation.

No 4 We showed in Section notes that about 92% of the power contained in a square pulse occurs in the main lobe of the frequency response,



In a radar system a pulse of high frequency carrier signal is transmitted. The frequency response is thus the same as a pulse but shifted to around the carrier frequency.



The bandwidth of the transmitter to ensure 90% power transmission is thus $\approx 2 \times \frac{1}{\tau}$.

$$\text{Bandwidth (B)} \geq 2 \times \frac{1}{\tau}$$

$$\text{In this case } B \geq 2 \times \frac{1}{200 \times 10^{-9}} = 10 \text{ MHz.}$$

Note a large bandwidth is needed to transmit this pulses.